



gr-harmonia: A Synchronization Toolkit for Software- defined Radios in a Distributed Network

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Outline



- Introduction
- Synchronization Procedure
- gr-harmonia: Blocks and Features
- Experimental Validation
- Conclusions and Future Work



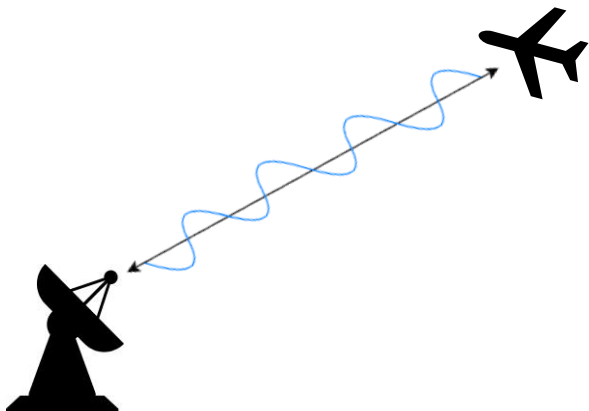
Outline



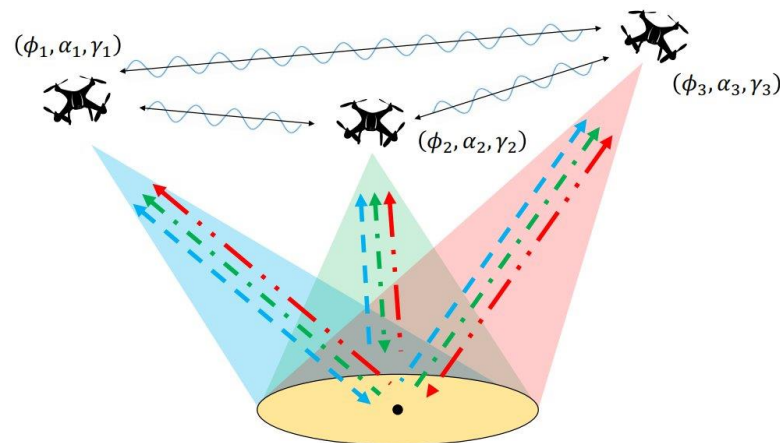
- **Introduction**
 - Motivation
 - Signal Model
- Synchronization Procedure
- gr-harmonia: Blocks and Features
- Experimental Results
- Conclusions and Future Work

Motivation

Traditional Monostatic Radar



Distributed Radar System



[1] R. H. Kenney and J. W. McDaniel, "Cooperative Navigation of Mobile Radar Sensors Using Time-of-Arrival Measurements and the Unscented Kalman Filter," in IEEE Transactions on Radar Systems, vol. 1, pp. 435–447, Aug. 2023.

[5] S. R. Mghabghab and J. A. Nanzer, "Open-Loop Distributed Beamforming Using Wireless Frequency Synchronization," IEEE Transactions on Microwave Theory and Techniques, vol. 69, no. 1, pp. 896–905, Jan. 2021.

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Introduction

Signal Model

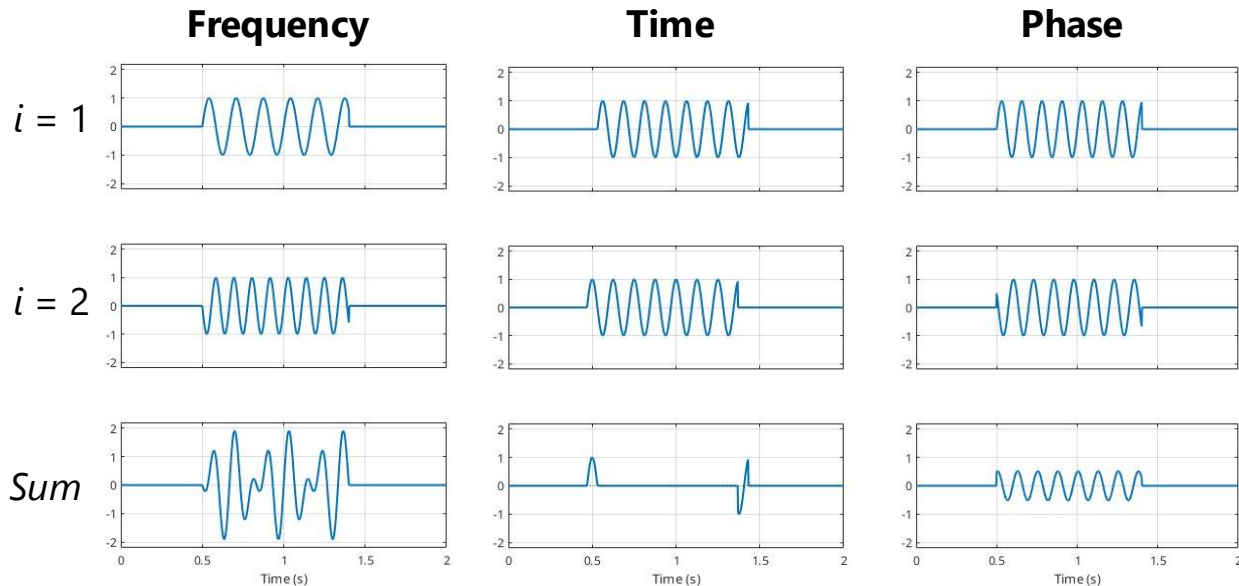
- Clock Drift (α)
- Clock Bias (ϕ)
- Carrier Phase ($\gamma^{\text{tx}}, \gamma^{\text{rx}}$)

Clock drift definition

$$\alpha_i = f_i^{\text{MO}} / f^{\text{MO}}$$

Local time on platform i

$$\tau_i = \alpha_i t + \phi_i$$



- Received signal can be expressed in terms of the receiver's local clock

$$r_{i,j}(\tau_i) = s_j \underbrace{\left(\frac{\alpha_j}{\alpha_i} \tau_i - \left(\frac{\alpha_j}{\alpha_i} \phi_i - \phi_j + \alpha_j \frac{R_{i,j}}{c} \right) \right)}_{\text{Baseband Signal}} \underbrace{\exp \left(j 2\pi f^c \left(\frac{\alpha_j}{\alpha_i} - 1 \right) \tau_i \right)}_{\text{Residual Carrier Frequency Error}} \underbrace{\exp \left(-j 2\pi f^c \left(\frac{\alpha_j}{\alpha_i} \phi_i - \phi_j + \alpha_j \frac{R_{i,j}}{c} \right) \right)}_{\text{Residual Clock Bias and Propagation Delay Phase}} \underbrace{\exp(j \gamma_{i,j}^{\text{err}})}_{\text{Carrier Phase Error}}$$



Outline



- Introduction
- **Synchronization Procedure**
 - Assumptions and Time-Division Multiple Access
 - Clock Drift Estimation and Compensation
 - Clock Bias and Carrier Phase Estimation and Compensation
- gr-harmonia: Blocks and Features
- Experimental Validation
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Synchronization Procedure

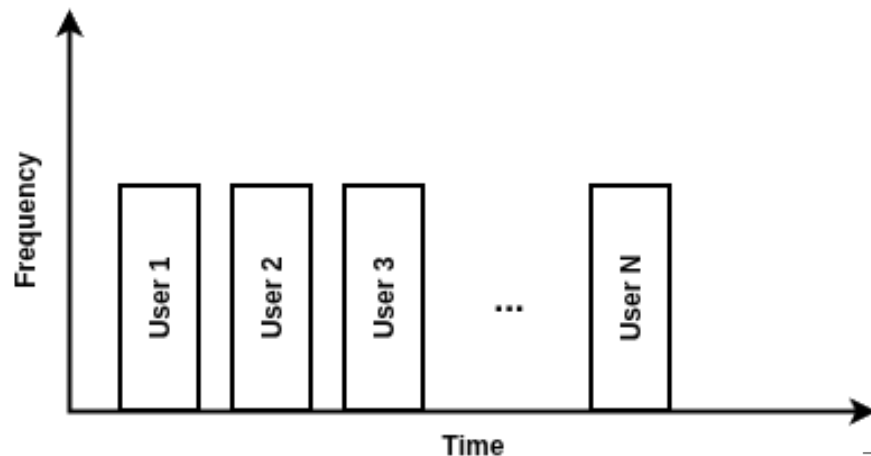


Assumptions

- Each distributed radar platform i has an independent main oscillator with a nominal frequency f^{MO} and a true frequency f_i^{MO}
- Each platform is stationary
- Coarse synchronization performed beforehand

Time-Division Multiple Access

$$\tau_j^{\text{tx}} = (j - 1)\Delta_{\text{TDMA}}, \quad j = 1, 2, \dots, N$$





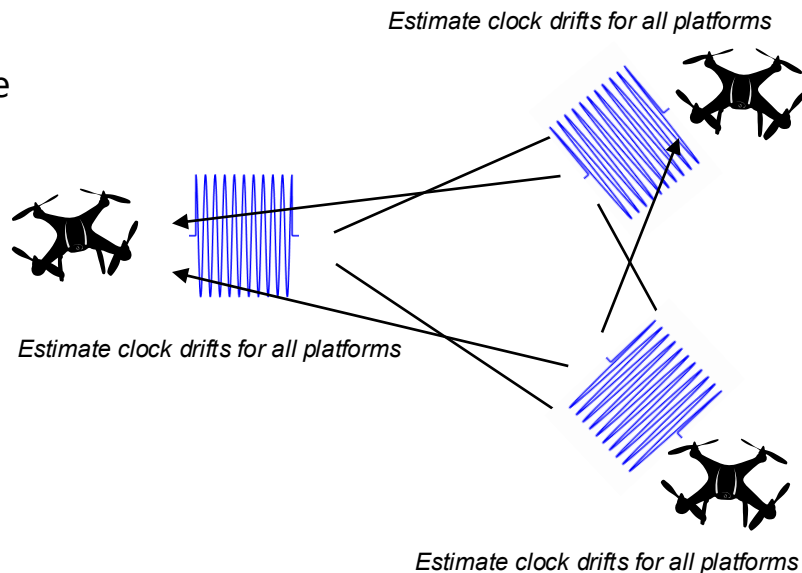
Synchronization Procedure



Clock Drift Estimation

1. Single-Tone Pulse Broadcasts

- Each platform **broadcasts** specific frequency pulse
- RX platforms **estimate** frequency from each TX
- Each platform **shares** estimated frequency
- Each platform **solves** equations for clock drifts
- Each platform **compensates** drift on TX and RX





Synchronization Procedure



Clock Drift Estimation

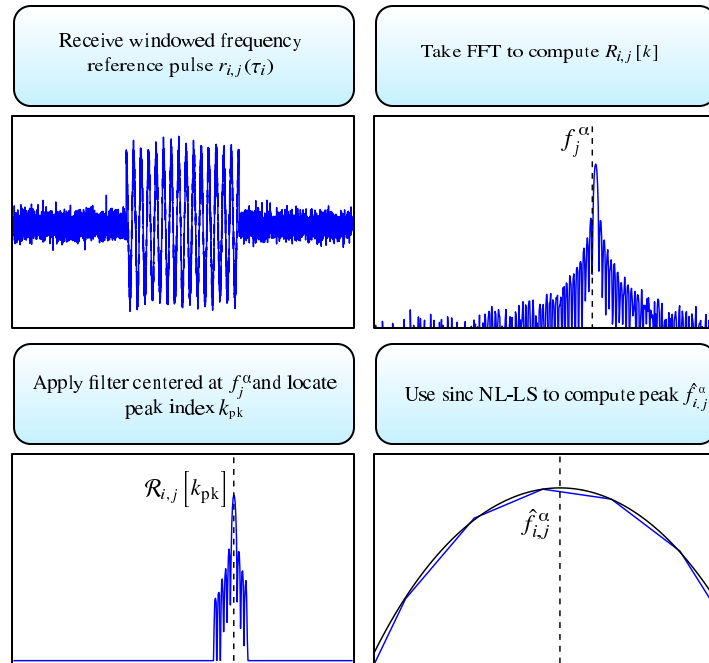
- Estimate frequency peak using sinc NL-LS [2]
 - Yields an estimate $\hat{f}_{i,j}^{\alpha}$
- All platforms exchange frequency estimates
- Estimates nominally related to clock drifts by

$$\hat{f}_{i,j}^{\alpha} = \left(\frac{\alpha_j}{\alpha_i} - 1 \right) f^c + \frac{\alpha_j}{\alpha_i} f_j^{\alpha}$$

- Rearrange to give

$$\left(\hat{f}_{i,j}^{\alpha} + f^c \right) \alpha_i - \left(f_j^{\alpha} + f^c \right) \alpha_j = 0$$

- Relative drifts – set one drift to unity (e.g., $\alpha_1 = 1$)
- Each platform uses linear LS to solve above equations for drift estimates $\hat{\alpha}_i$





Synchronization Procedure

Clock Drift Compensation

- On transmit: recompute baseband signal w/ new time axis and apply carrier correction

$$s'_j(\tau_j) = s_j\left(\frac{\tau_j}{\hat{\alpha}_j}\right) \exp\left(j 2\pi f^c \left(\frac{1}{\hat{\alpha}_j} - 1\right) \tau_j\right)$$

- On receive: scale RX signal w/ new time axis $\tau'_i = \tau_i / \hat{\alpha}_i$ and apply carrier correction

$$\begin{aligned} r_{i,j}^b(\tau'_i) &= r_{i,j}(\tau'_i) \exp(j 2\pi f^c (\hat{\alpha}_i - 1) \tau'_i) \\ &= s_j\left(\tau'_i - \left(\frac{1}{\hat{\alpha}_i} \phi_i - \frac{1}{\hat{\alpha}_j} \phi_j + \frac{\alpha_j}{\hat{\alpha}_j} \frac{R_{i,j}}{c}\right)\right) \exp\left(-j 2\pi f^c \left(\frac{1}{\hat{\alpha}_i} \phi_i - \frac{1}{\hat{\alpha}_j} \phi_j + \frac{\alpha_j}{\hat{\alpha}_j} \frac{R_{i,j}}{c}\right)\right) \exp(j \gamma_{i,j}^{\text{err}}) \end{aligned}$$

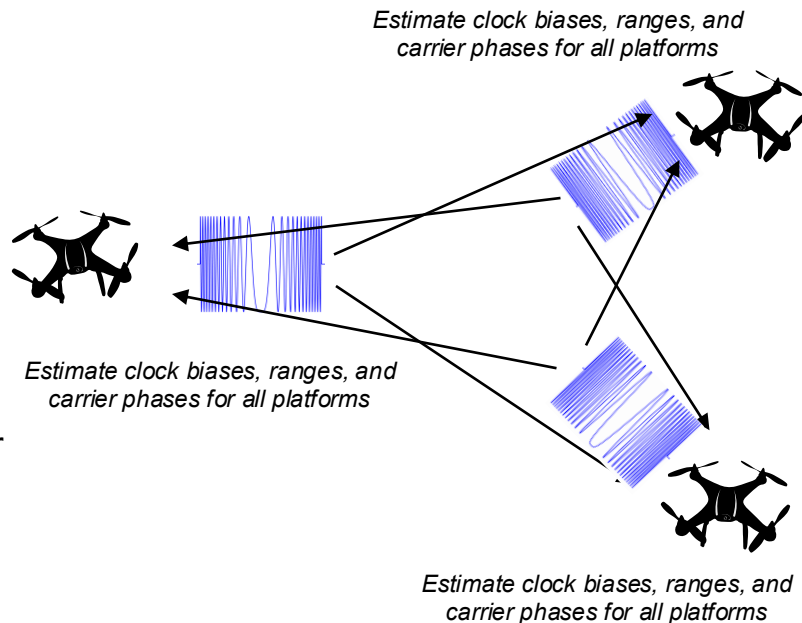


Synchronization Procedure



Clock Bias and Carrier Phase Estimation

1. Linear Frequency Modulated (LFM) Pulse Broadcasts
 - a) Each platform **broadcasts** ranging waveform
 - b) Each platform **estimates** time-delay and phase
 - c) Platforms **share** time-delay and phase estimates
 - d) Each platform **solves** for range, clock biases, and carrier phases
 - e) Each platform **compensates** clock bias and carrier phase on TX and RX





Synchronization Procedure

Clock Bias and Carrier Phase Estimation

- Assumes clock drifts now compensated – works identically to procedure in [2]
- Pulse-compression waveforms broadcast amongst platforms, and time-delays estimated

$$m_{i,j} = \frac{1}{\hat{\alpha}_i} \phi_i - \frac{1}{\hat{\alpha}_j} \phi_j + \frac{\alpha_j}{\hat{\alpha}_j} \frac{R_{i,j}}{c}$$

- Share time-delay estimates and phase at pulse-compression peak with all other platforms
- Exploit signal symmetry to compute range and bias differences

$$\hat{R}_{i,j} = \frac{\alpha_j}{\hat{\alpha}_j} R_{i,j} = \frac{c(m_{i,j} + m_{j,i})}{2}$$

$$\hat{\phi}_{i,j} = \frac{1}{\hat{\alpha}_i} \phi_i - \frac{1}{\hat{\alpha}_j} \phi_j = \frac{m_{i,j} - m_{j,i}}{2}$$

- Compute network average of clock biases and estimates of carrier phase differences

$$\tilde{\phi}_i = \frac{1}{N} \sum_j \hat{\phi}_{i,j}$$

$$\hat{\gamma}_{i,j}^{\text{err}} = \angle d_{i,j}(m_{i,j}) + 2\pi f^c \left(\frac{\hat{R}_{i,j}}{c} + \hat{\phi}_{i,j} \right)$$



Synchronization Procedure



Clock Bias and Carrier Phase Compensation

- For all future TX signals at platform i :
 - Apply fractional delay of $\tilde{\phi}_i$ to correct bias in delay
 - Apply phase correction of $\exp(-j 2\pi f^c \tilde{\phi}_i)$ to correct bias phase
 - Apply phase correction of $\exp(-j \hat{\gamma}_i^{\text{TX}})$ to correct carrier phase
- For all future RX signals at platform i :
 - Apply fractional delay of $-\tilde{\phi}_i$ to correct bias in delay
 - Apply phase correction of $\exp(j 2\pi f^c \tilde{\phi}_i)$ to correct bias phase
 - Apply phase correction of $\exp(j \hat{\gamma}_i^{\text{RX}})$ to correct carrier phase



Outline



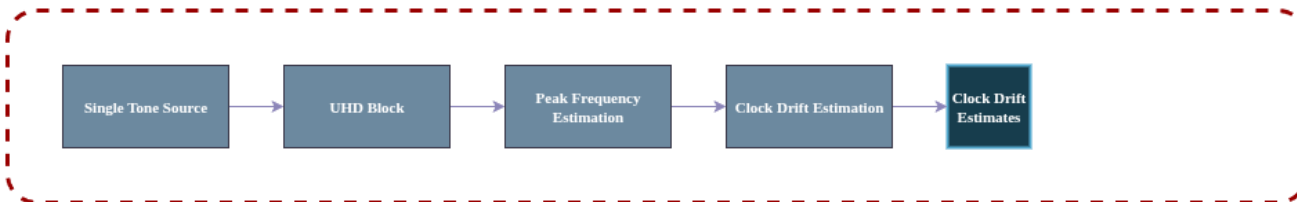
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- **gr-harmonia: Blocks and Features**
 - Waveform Generation
 - USRP Interfacing and TDMA
 - Clock Drift Estimation
 - Clock Bias and Carrier Phase Estimation
 - Waveform Compensation
- Experimental Validation
- Conclusions and Future Work



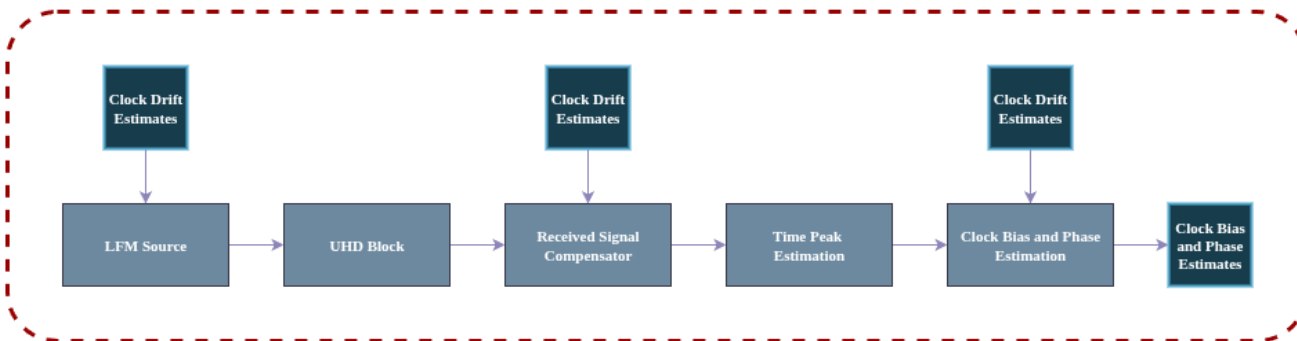
gr-harmonia: Blocks and Features

- All blocks operate in the message domain and exchange **protocol data units (PDUs)** via message ports.

Clock Drift Estimation



Clock Bias and Carrier Phase Estimation



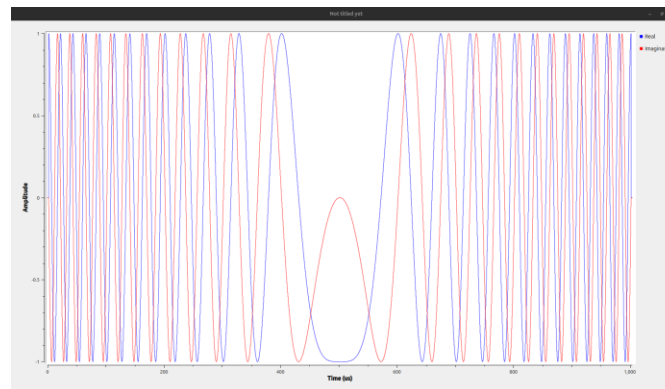
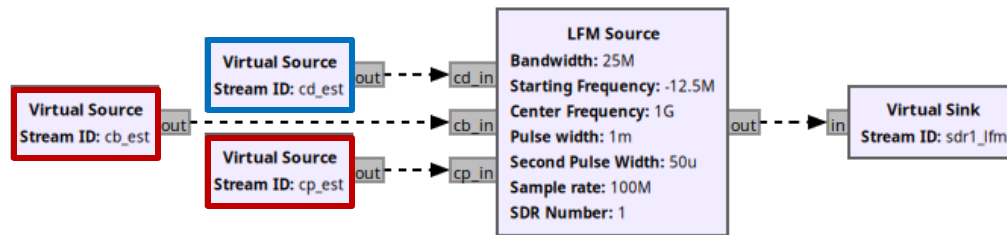
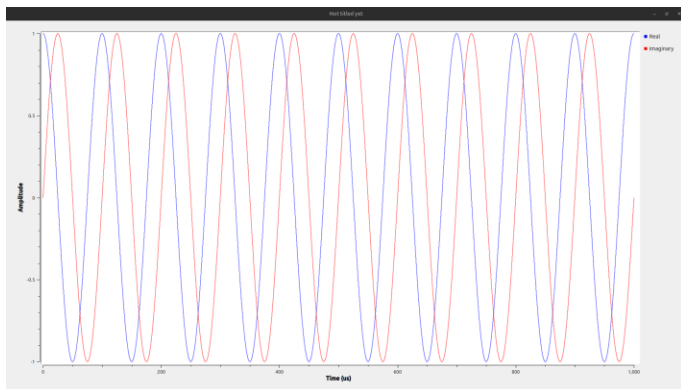
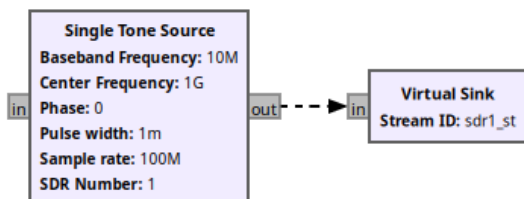


gr-harmonia: Blocks and Features



Waveform Generation

- **Single-Tone** and **LFM** Block



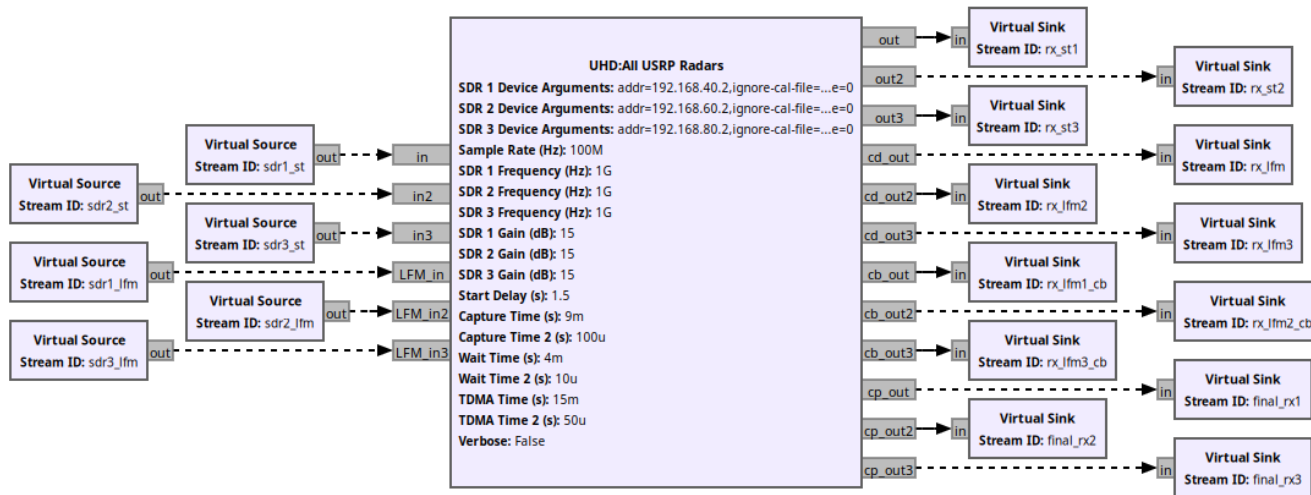


gr-harmonia: Blocks and Features



UHD Interfacing and TDMA

- **UHD: ALL USRP** Block



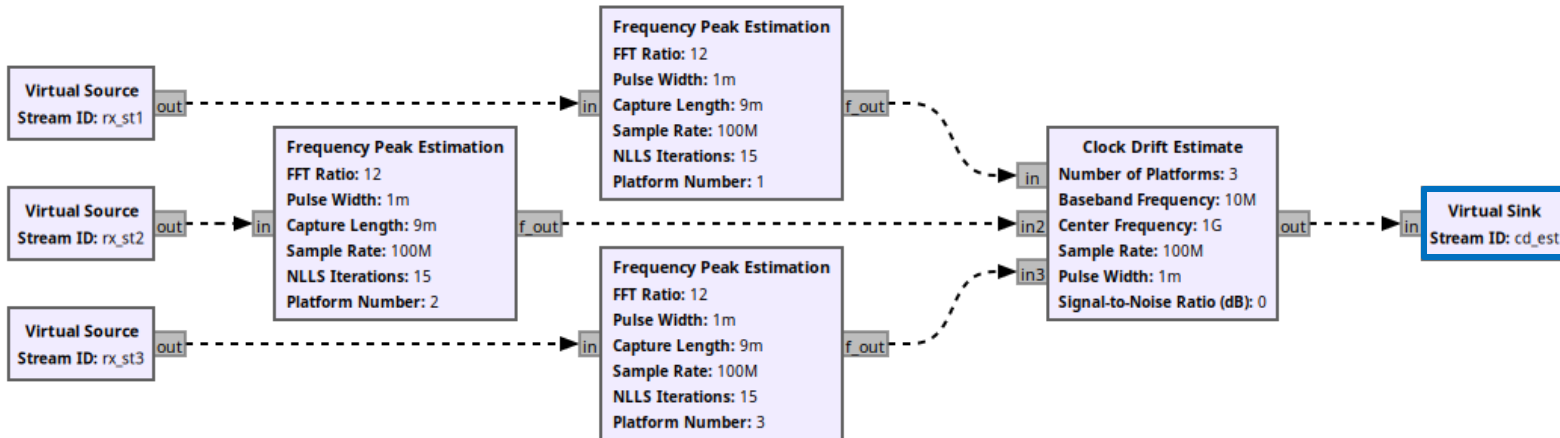


gr-harmonia: Blocks and Features



Clock Drift Estimation

- **Frequency Peak Estimation** and **Clock Drift Estimation** Block

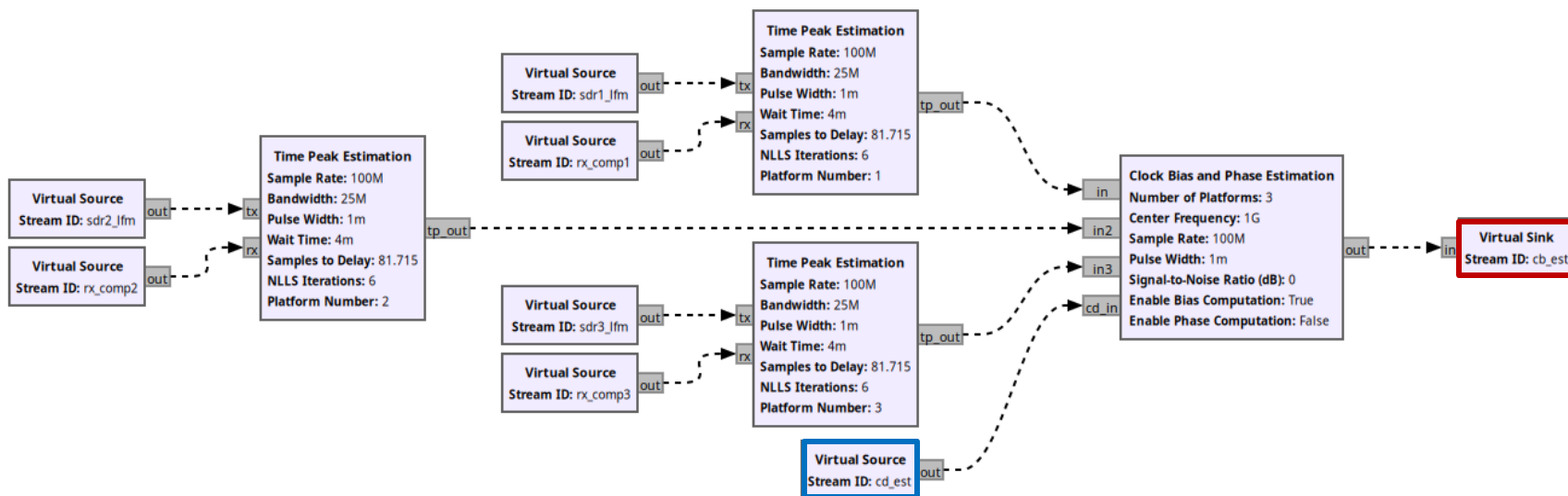




gr-harmonia: Blocks and Features

Clock Bias and Carrier Phase Estimation

- **Time Peak Estimation** and **Clock Bias and Carrier Phase Estimation** Block



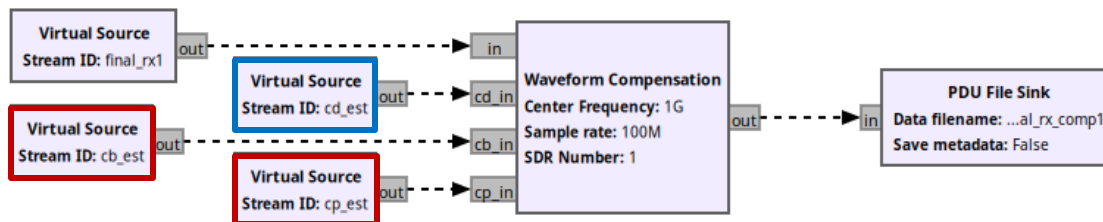


gr-harmonia: Blocks and Features



Waveform Compensation

- **Compensation Block**





Outline



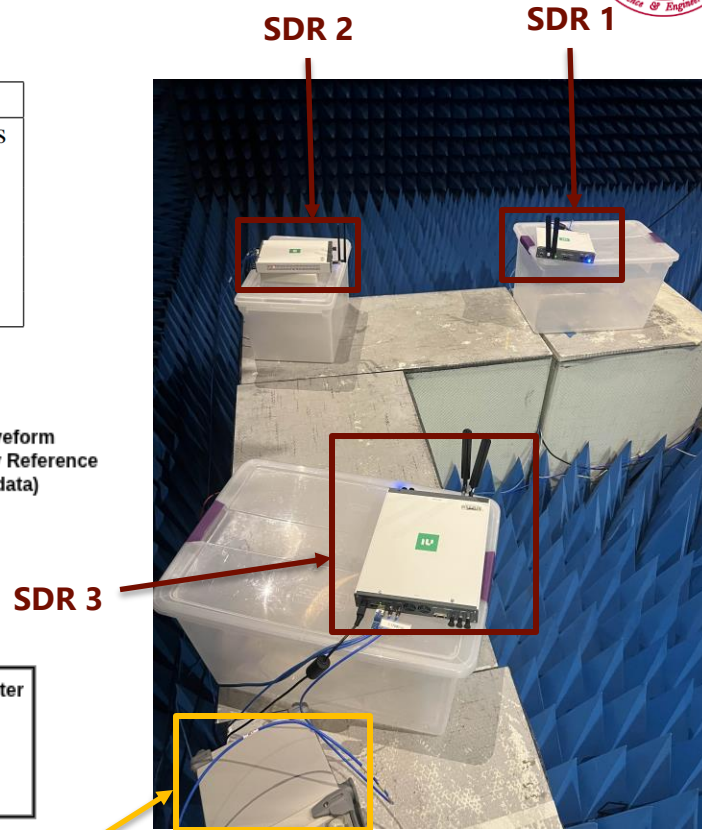
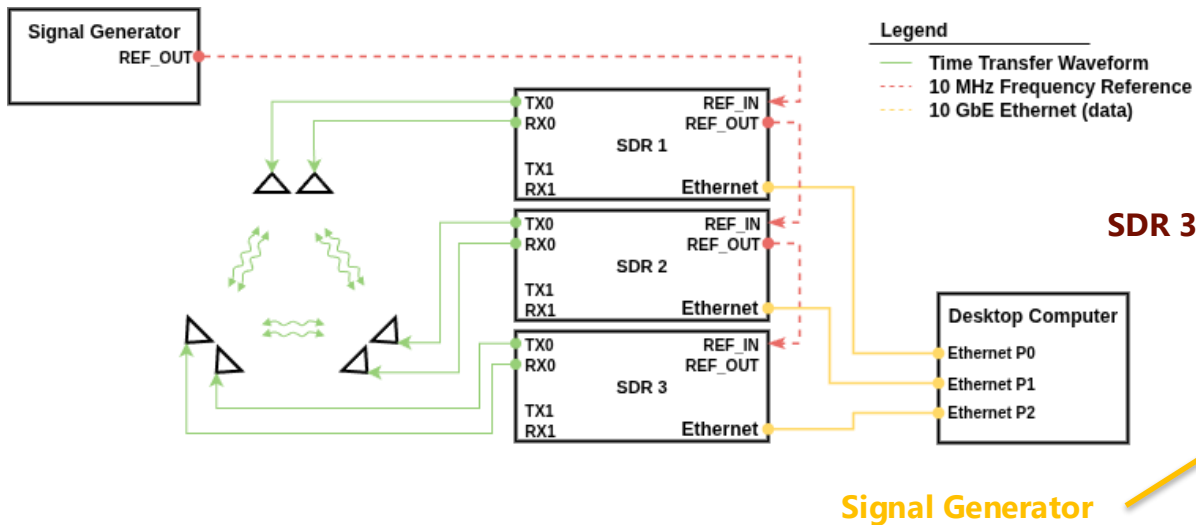
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Experimental Validation

Configuration

- Over-the-air experiments
- Ettus X310s with UBX 160 daughterboards
- Monopole antennas

Parameters	Value
Sample rate	100 MS/s
Center frequency	1 GHz
Baseband frequency	10 MHz
Bandwidth	25 MHz
Pulse width	1 ms
Capture length	9 ms





Experimental Validation



Experimental Results

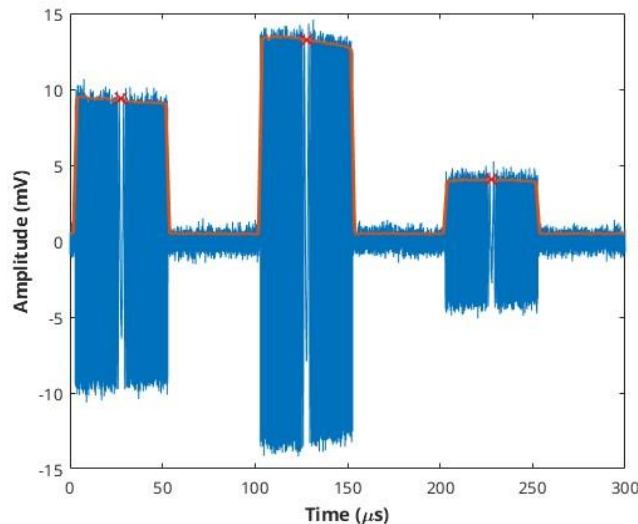
- External 10 MHz reference clock
 - Measure the performance of clock drift

SDR	Standard Deviation
2	7.17 ppb
3	6.77 ppb

- Internal reference clock and external reference clock assuming ideal clock drift estimation
 - Measure the performance of clock bias and carrier phase

Test	Dev 1 → Dev 2	Dev 1 → Dev 3	Dev 2 → Dev 3
	Standard Deviation		
Internal Clock	26.18 cm (0.87 ns)	45.73 cm (1.52 ns)	61.57 cm (2.05 ns)
External Clock	3.22 cm (0.10 ns)	3.92 cm (0.13 ns)	4.48 cm (0.15 ns)

Coherent Gain: 98.62%





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Conclusion and Future Work

Conclusion

- Implemented a distributed SDR synchronization module
- Characterized clock drift, clock bias, and carrier phase estimation
- Reduce clock drift to ~ 7 ppb
- Achieved a maximum of 98.62% gain and reduced clock bias to 3-5 cm with an external 10 MHz reference

Future Work

- Improve documentation of module
- Add UHD block for multiple host implementations
- Incorporate SNR estimation, weighted least squares (WLS) estimation, and Kalman filtering to improve module performance

Thank you!





References

- [1] R. H. Kenney and J. W. McDaniel, "Cooperative Navigation of Mobile Radar Sensors Using Time-of-Arrival Measurements and the Unscented Kalman Filter," in IEEE Transactions on Radar Systems, vol. 1, pp. 435–447, Aug. 2023.
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